

Achievable Rate Analysis in Network-Coded Cooperative Communications with Multiple Relay Nodes

Sushant Sharma* Yi Shi* Y. Thomas Hou* Sastry Kompella†

* Virginia Polytechnic Institute and State University, Blacksburg, VA, USA

† U.S. Naval Research Laboratory, Washington D.C., USA

Abstract—Network-coded cooperative communications (NC-CC) refers to the use of network coding (NC) in cooperative communications (CC). Prior studies have shown that NC has the potential to improve the performance of CC when there are multiple sessions in the wireless network. These studies were done for the case when multiple sessions are sharing a single relay node. However, how NC-CC behaves when multiple relay nodes are employed remains an open problem. In this paper, we explore this problem by analyzing the achievable rate of each session in this setting. We develop closed form formulas for the mutual information and the achievable data rate for each session and show that prior results for a single relay is a special case of our result. Our findings in this paper offer an important building block on the theory of NC-CC.

I. INTRODUCTION

Cooperative communications (CC) is an important technique to improve the performance of a wireless network [5]. Unlike MIMO, which requires the use of multiple antennas at each node, CC only employs a single antenna at each node and exploits diversity by cooperating with antennas on other nodes in the network. There has been extensive research at the physical layer that exploits cooperations among distributed antennas [1], [3], [6].

Recently, it was found that network coding (NC) can further improve the performance of CC by combining data streams at a relay node [2], [9], [10], [13], [14]. This application of NC in CC is called network-coded CC (or NC-CC). NC has been shown to improve the outage probability ([2], [9]), packet error rates ([13]), and data rates ([10], [14]) for CC. These studies were done for the case when multiple sessions are sharing a single relay node. However, how NC-CC behaves when multiple relay nodes are employed remains an open problem.

In this paper, we study NC-CC when there are multiple relay nodes. Our goal is to develop closed form formulas for the mutual information and the achievable data rate for each session. We consider the case of analog network coding (ANC) [4] and amplify-and-forward (AF) CC [7] at each relay node. Through an in-depth analysis, we derive the mathematical equations for mutual information and achievable data rate for each session and show that prior results for a single relay is a special case of our result. Our findings in this paper offer an important building block on the theory of NC-CC. To demonstrate the application of our theoretical result, we apply it in a numerical study to understand the impact on a session's achievable rate when different sets of relay nodes are employed.

The remainder of this paper is organized as follows. In Section II, we review state-of-the-art result on NC-CC when

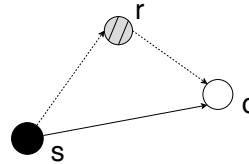


Fig. 1. A three-node relay channel for CC.

only a single relay node is employed. In Section III, we study NC-CC with multiple relay nodes and develop formulas for the mutual information and achievable data rate of each session. Section IV illustrates our results in a numerical study. Section V concludes this paper. Table I lists all notation used in this paper.

II. BACKGROUND

As a simple example, Fig. 1 shows a three-node relay channel for CC. In this example, source node s intends to transmit to destination node d and will exploit relay node r for possible performance improvement. Assume the time frame for transmission is divided into two time slots. In the first time slot, s transmits to d , and is overheard by the relay node r . Relay node r then amplifies the received signal, and then retransmits the amplified signal in the second time slot. The destination node d can now combine the two copies of the same signal coming from two different paths. This cooperative relay channel in Fig. 1 can be treated as a single-input two-output complex Gaussian channel [7] and the achievable data rate between s and d in this channel is given as follows:

$$C_{cc}(s, r, d) = \frac{W}{2} \log_2 \left(1 + \text{SNR}_{sd} + \frac{\text{SNR}_{sr} \cdot \text{SNR}_{rd}}{1 + \text{SNR}_{sr} + \text{SNR}_{rd}} \right), \quad (1)$$

where SNR_{uv} is the signal-to-noise ratio at the receiver v when node u transmits, and is given by $\text{SNR}_{uv} = \frac{|h_{uv}|^2 P_u}{\sigma_v^2}$, σ_v^2 is the variance of background noise at node v , h_{uv} is the gain of the channel from node u to node v , P_u is the power at which node u transmits signals, and W is the channel bandwidth.

As for comparison, when CC is not used, i.e., s transmits to d without using r (so-called direct transmission), the achievable rate from s to d is given as:

$$C_d(s, \emptyset, d) = W \log_2 (1 + \text{SNR}_{sd}), \quad (2)$$

where \emptyset denotes that no relay node is used. It has been shown in [7] and [11] that CC has the potential to increase achievable

TABLE I
NOTATION

Symbol	Definition
α_r	Amplification factor at relay node r
σ_v^2	Variance of background noise at node v
$\sigma_{z_v}^2$	Variance of ANC noise at node v
h_{uv}	Effect of path-loss, shadowing, and fading from node u to node v
z_v	White Gaussian background noise at node v
z_v^{ANC}	ANC noise at node v
$C_{\text{NC-CC}}(s_i, \mathcal{S}, \mathcal{R}, d_i)$	Achievable rate for session $(s_i, d_i) : s_i \in \mathcal{S}$ employs NC-CC using the set of relays \mathcal{R}
$I_{\text{NC-CC}}(s_i, \mathcal{S}, \mathcal{R}, d_i)$	Mutual information for session $(s_i, d_i) : s_i \in \mathcal{S}$ that employs relay group \mathcal{R}
P_u	Transmission power at node u
\mathcal{R}	The set of all relay nodes in the network
\mathcal{S}	The set of all source nodes in the network
SNR_{uv}	The signal to noise ratio between nodes u and v
W	Total bandwidth available in the network

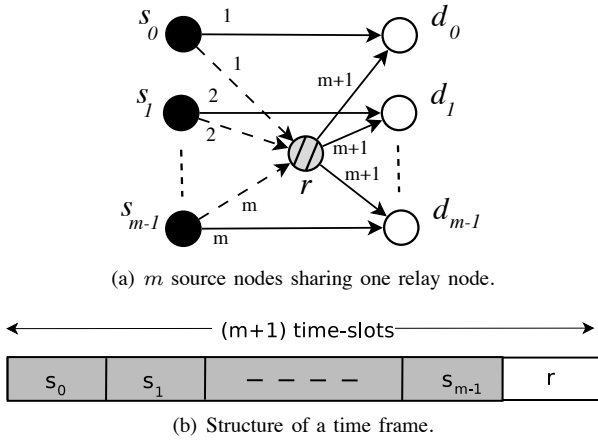


Fig. 2. NC-CC with a single relay node.

rate over direct transmission, depending on the location of relay node and its channel statistics.

When there are m source-destination sessions sharing a single relay node (as shown in Fig. 2), one can employ NC to combine the signals from the m sources at the relay node and then forward the combined signal to all the destination nodes, i.e., NC-CC. Here, a time frame is divided into $(m+1)$ time slots (see Fig. 2(b)), with each of the first m time slots being used for transmission by one of the source nodes. Again, each of such transmissions is received by its corresponding destination node and overheard by the relay node. The relay node will then apply NC to combine all the m received analog signals. The combined signal is then amplified and forwarded by the relay node to all destination nodes in the $(m+1)$ -th time slot. The achievable rate for a session under NC-CC is given by [10]:

$$C_{\text{NC-CC}}(s_i, \mathcal{S}, r, d_i) = \frac{W}{m+1} \log_2 \left(1 + \text{SNR}_{s_i d_i} + \frac{\text{SNR}_{s_i r} \text{SNR}_{r d_i}}{|\mathcal{S}_r| \frac{\sigma_{z_{d_i}}^2}{\sigma_{d_i}^2} + \text{SNR}_{r d_i} + \frac{\sigma_{z_{d_i}}^2}{\sigma_{d_i}^2} \sum_{s_j \in \mathcal{S}} \text{SNR}_{s_j r}} \right), \quad (3)$$

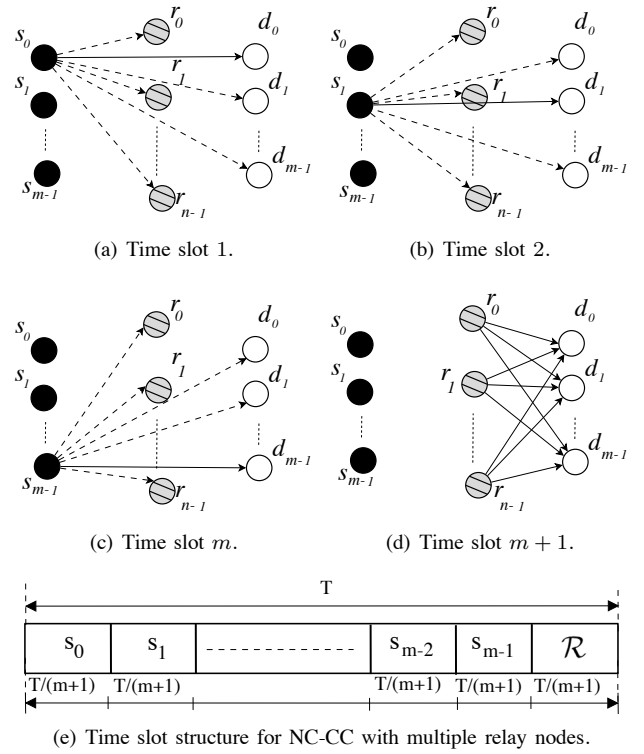


Fig. 3. Data transfer under multi-relay NC-CC in $m+1$ time-slots.

where $\mathcal{S} = \{s_0, s_1, \dots, s_{m-1}\}$ is the set of all the source nodes, and $\sigma_{z_{d_i}}^2$ is the noise at destination d_i due to ANC, and is given by

$$\sigma_{z_{d_i}}^2 = \sigma_{d_i}^2 + (|\mathcal{S}| - 1) (\alpha_r h_{r d_i})^2 \sigma_r^2 + \sigma_{d_i}^2 \sum_{s_j \in \mathcal{S}, s_j \neq s_i} \left(\frac{\alpha_r h_{s_j r} h_{r d_i}}{h_{s_j d_i}} \right)^2, \quad (4)$$

where α_r is the amplification factor for AF CC at relay node r and is given by

$$\alpha_r^2 = \frac{P_r}{|\mathcal{S}| \sigma_r^2 + \sum_{s_i \in \mathcal{S}} P_{s_i} |h_{s_i r}|^2}. \quad (5)$$

III. NC-CC WITH MULTIPLE RELAY NODES

The NC-CC model in Fig. 2 employs only a single relay node. In this section, we study the general NC-CC model with multiple relay nodes, and derive mutual information and achievable data rate for each session.

We consider that there are m sessions in the network (denoted as $\mathcal{S} = \{s_0, s_1, \dots, s_{m-1}\}$) that are sharing n relay nodes with NC-CC. This general multi-session multi-relay NC-CC model is illustrated in Fig. 3. In this figure, the source node of each session transmits in the first m time slots, and each transmission is received by the destination nodes as well as the n relay nodes. After the m transmissions, all the relay nodes will apply NC to combine the signals received during the m time slots. Figure 3(d) shows that the combined signal is amplified and simultaneously transmitted by all the relay nodes

in the $(m+1)$ -th time slot. The coherent signal reception at destination nodes can be facilitated by appropriate transmitter synchronization [8]. Figure 3(e) shows the time-slot structure for this general multi-session multi-relay model. Every session is allotted a time slot duration of t . Thus, the time slot duration available to each source node is $\frac{T}{m+1} = \frac{|\mathcal{S}|t}{|\mathcal{S}|+1}$. Therefore, the achievable rate for a session (s_i, d_i) can be written as:

$$\begin{aligned} C_{\text{NC,CC}}(s_i, \mathcal{S}, \mathcal{R}, d_i) &= W \cdot \frac{\left(\frac{|\mathcal{S}|t}{|\mathcal{S}|+1}\right)}{|\mathcal{S}|t} I_{\text{NC,CC}}(s_i, \mathcal{S}, \mathcal{R}, d_i) \\ &= \frac{W}{|\mathcal{S}|+1} I_{\text{NC,CC}}(s_i, \mathcal{S}, \mathcal{R}, d_i), \end{aligned} \quad (6)$$

where $I_{\text{NC,CC}}(s_i, \mathcal{S}, \mathcal{R}, d_i)$ is the mutual information between the source s_i and its destination node d_i .

Now we derive the mutual information $I_{\text{NC,CC}}(s_i, \mathcal{S}, \mathcal{R}, d_i)$. First, the signal transmitted by source s_i and received by destination d_i can be written as:

$$y_{s_i d_i} = h_{s_i d_i} x_i + z_{d_i}. \quad (7)$$

Next, the signal transmitted by source s_i and received by relay r_j can be written as:

$$y_{s_i r_j} = h_{s_i r_j} x_i + z_{r_j}. \quad (8)$$

Finally, the simultaneous signals transmitted by all the relay nodes and received by destination node d_i in the $(m+1)$ -th time slot can be written as:

$$\begin{aligned} y_{\mathcal{S} \mathcal{R} d_i} &= \sum_{r_j \in \mathcal{R}} \left(\alpha_{r_j} h_{r_j d_i} \sum_{s_k \in \mathcal{S}} y_{s_k r_j} \right) + z_{d_i} \\ &= \sum_{r_j \in \mathcal{R}} \left(\alpha_{r_j} h_{r_j d_i} \sum_{s_k \in \mathcal{S}} (h_{s_k r_j} x_k + z_{r_j}) \right) + z_{d_i}, \end{aligned} \quad (9)$$

where the value of amplification factor at r_j is:

$$\alpha_{r_j}^2 = \frac{P_{r_j}}{\sum_{s_k \in \mathcal{S}} (\sigma_{r_j}^2 + P_{s_k} |h_{s_k r_j}|^2)}.$$

A destination node d_i has to extract signal for x_i from the combined signal in (9). To extract the desired signal, a destination node d_i can subtract the signals received from other source nodes in the initial m time slots from this combined signal. The signals received by destination node d_i from other source nodes during the first m time slots can be written as:

$$y_{s_k d_i} = h_{s_k d_i} x_k + z_{d_i}, \quad (s_k \in \mathcal{S}, s_k \neq s_i). \quad (10)$$

To remove the signal x_k from the combined signal in (9), destination node d_i can multiply the overheard signal for x_k in (10) by $\frac{h_{s_k d_i}}{h_{s_k r_j}}$, and then subtract this product from (9). When the signals for all x_k 's are removed from (9), the copy of the desired signal extracted by d_i can

be written as:

$$\begin{aligned} y_{\mathcal{S} \mathcal{R} d_i} &= x_i \sum_{r_j \in \mathcal{R}} h_{s_i r_j} \alpha_{r_j} h_{r_j d_i} + \sum_{s_k \in \mathcal{S}} \sum_{r_j \in \mathcal{R}} \alpha_{r_j} h_{r_j d_i} z_{r_j} \\ &\quad + z_{d_i} - \sum_{s_k \in \mathcal{S}} \frac{\sum_{r_j \in \mathcal{R}} h_{s_k r_j} \alpha_{r_j} h_{r_j d_i}}{h_{s_k d_i}} \\ &= \sum_{r_j \in \mathcal{R}} \alpha_{r_j} h_{r_j d_i} y_{s_i r_j} + z_{d_i} + \sum_{s_k \in \mathcal{S}} \sum_{r_j \in \mathcal{R}} \alpha_{r_j} h_{r_j d_i} z_{r_j} \\ &\quad - \sum_{s_k \in \mathcal{S}} \frac{\sum_{r_j \in \mathcal{R}} h_{s_k r_j} \alpha_{r_j} h_{r_j d_i}}{h_{s_k d_i}} z_{d_i}. \end{aligned} \quad (11)$$

Due to the use of ANC and multiple relay nodes, we find in (11) that in addition to z_{d_i} , there are some new noise terms in the signal extracted by d_i . This new noise is called the ANC noise, and is written as:

$$\begin{aligned} z_{d_i}^{\text{ANC}} &= z_{d_i} + \sum_{s_k \in \mathcal{S}} \sum_{r_j \in \mathcal{R}} \alpha_{r_j} h_{r_j d_i} z_{r_j} \\ &\quad - \sum_{s_k \in \mathcal{S}} \frac{\sum_{r_j \in \mathcal{R}} h_{s_k r_j} \alpha_{r_j} h_{r_j d_i}}{h_{s_k d_i}} z_{d_i}. \end{aligned}$$

We can now write the variance of ANC noise as:

$$\begin{aligned} \sigma_{z_{d_i}^{\text{ANC}}}^2 &= \sigma_{d_i}^2 + (|\mathcal{S}| - 1) \sum_{r_j \in \mathcal{R}} (\alpha_{r_j} h_{r_j d_i})^2 \sigma_{r_j}^2 \\ &\quad + \sum_{s_k \in \mathcal{S}} \sum_{r_j \in \mathcal{R}} \left(\frac{h_{s_k r_j} \alpha_{r_j} h_{r_j d_i}}{h_{s_k d_i}} \right)^2 \sigma_{d_i}^2. \end{aligned} \quad (12)$$

From (12), we can make two important observations: (i) the variance in ANC noise increases monotonically as the group size of sessions sharing the same set of relays increase; and (ii) the variance in ANC noise increases monotonically with the size of the set of relays.

We can now write (7) and (11) in a compact matrix form as follows:

$$\mathbf{Y} = \mathbf{H}x_i + \mathbf{B}\mathbf{Z},$$

where

$$\mathbf{Y} = \begin{bmatrix} y_{s_i d_i} \\ y_{\mathcal{S} \mathcal{R} d_i} \end{bmatrix}, \quad \mathbf{H} = \begin{bmatrix} h_{s_i d_i} \\ \sum_{r_j \in \mathcal{R}} h_{s_i r_j} \alpha_{r_j} h_{r_j d_i} \end{bmatrix},$$

$$\mathbf{B} = \begin{bmatrix} 0 & 0 & \cdots & 0 & 1 & 0 \\ \alpha_r h_{r_0 d_i} & \alpha_r h_{r_1 d_i} & \cdots & \alpha_r h_{r_{n-1} d_i} & 0 & 1 \end{bmatrix}, \text{ and}$$

$$\mathbf{Z} = \begin{bmatrix} z_{r_0} \\ z_{r_1} \\ \vdots \\ z_{r_{n-1}} \\ z_{d_i} \\ z_{d_i}^{\text{ANC}} \end{bmatrix}.$$

We can model the above channel from $s_i \rightarrow d_i$ and from $s_i \rightarrow \mathcal{R} \rightarrow d_i$ as a one-input two-output complex Gaussian

channel [7]. The mutual information between s_i and d_i can be written as:

$$I_{\text{NC-CC}}(s_i, \mathcal{S}, \mathcal{R}, d_i) = \log \det \left\{ \mathbf{I} + (P_{s_i} \mathbf{H} \mathbf{H}^\dagger) (\mathbf{B} \mathbf{E} [\mathbf{Z} \mathbf{Z}^\dagger] \mathbf{B}^\dagger)^{-1} \right\}, \quad (13)$$

where

$$\mathbf{E} [\mathbf{Z} \mathbf{Z}^\dagger] = \begin{bmatrix} \sigma_{r_0}^2 & 0 & \cdots & 0 & 0 & 0 \\ 0 & \sigma_{r_1}^2 & \cdots & 0 & 0 & 0 \\ 0 & 0 & \ddots & 0 & 0 & 0 \\ 0 & 0 & \cdots & \sigma_{r_{n-1}}^2 & 0 & 0 \\ 0 & 0 & \cdots & 0 & \sigma_{d_i}^2 & 0 \\ 0 & 0 & \cdots & 0 & 0 & \sigma_{z_{d_i}^{\text{ANC}}}^2 \end{bmatrix}.$$

Expanding (13) gives us the value of mutual information between s_i and d_i as follows:

$$I_{\text{NC-CC}}(s_i, \mathcal{S}, \mathcal{R}, d_i) = \log_2 \left[1 + \frac{P_{s_i} |h_{s_i d_i}|^2}{\sigma_{d_i}^2} + \frac{P_{s_i} \left(\sum_{r_j \in \mathcal{R}} h_{s_i r_j} \alpha_{r_j} h_{r_j d_i} \right)^2}{\sigma_{z_{d_i}^{\text{ANC}}}^2 + \sum_{r_j \in \mathcal{R}} h_{r_j d_i}^2 \alpha_{r_j}^2 \sigma_{r_j}^2} \right],$$

which can be rewritten as:

$$I_{\text{NC-CC}}(s_i, \mathcal{S}, \mathcal{R}, d_i) = \log_2 \left[1 + \text{SNR}_{s_i d_i} + \frac{\left(\sum_{r_j \in \mathcal{R}} \sqrt{\frac{\text{SNR}_{r_j d_i} \text{SNR}_{s_i r_j}}{|\mathcal{S}| + \sum_{s_k \in \mathcal{S}} \text{SNR}_{s_k r_j}}} \right)^2}{\frac{\sigma_{z_{d_i}^{\text{ANC}}}^2}{\sigma_{d_i}^2} + \sum_{r_j \in \mathcal{R}} \left(\frac{\text{SNR}_{r_j d_i}}{|\mathcal{S}| + \sum_{s_k \in \mathcal{S}} \text{SNR}_{s_k r_j}} \right)} \right]. \quad (14)$$

From (14), we can see that the value of mutual information depends directly on the individual relay nodes in \mathcal{R} and the value of ANC noise. We can make two important observations by taking a close look at (14). First, a given session (s_i, d_i) performing NC-CC with a group of relays (r_j 's) would prefer these relays to stay as close as possible to s_i . Intuitively, this is true because the signal that r_j 's will receive from s_i will have smaller noise component (or larger SNR) due to proximity to s_i . A mathematical explanation of this is that the term $\text{SNR}_{s_i r_j}$ is in the denominator of (14). Second, s_i would also prefer to have other s_k 's in the session group to be far away from the relay nodes. This is because when the distance between the other source nodes (i.e., other s_k 's) and the relay nodes increases, the value of channel gains between other s_k 's and the relays become small, which will help reduce the ANC noise value at destination d_i . This can be observed in (12) where the value of ANC noise decreases monotonically with the decrease in the value of $h_{s_k r_j}$.

IV. NUMERICAL RESULTS

The goal of this section is to apply the achievable rate equations (6) and (14) in a numerical study to calculate the data rates of each session when different set of relay nodes are employed.

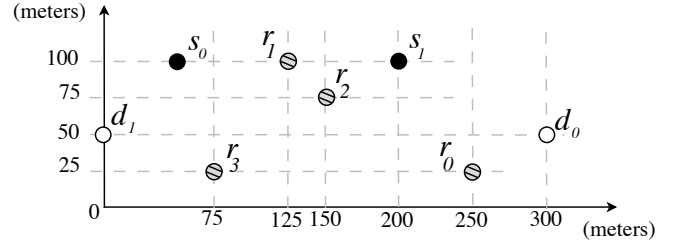


Fig. 4. Two-session four-relay network.

TABLE II
DATA RATES UNDER NC-CC WITH DIFFERENT SETS OF RELAY NODES.

\mathcal{R}	Data Rate (Mbps)		Total (Mbps)
	(s_0, d_0)	(s_1, d_1)	
$\mathcal{R}_0 = \emptyset$	16.70	25.83	42.53
$\mathcal{R}_1 = \{r_0\}$	13.65	20.01	33.66
$\mathcal{R}_2 = \{r_1\}$	18.48	27.44	45.92
$\mathcal{R}_3 = \{r_2\}$	13.81	28.80	42.61
$\mathcal{R}_4 = \{r_3\}$	17.50	23.14	40.64
$\mathcal{R}_5 = \{r_0, r_1\}$	15.45	31.78	47.23
$\mathcal{R}_6 = \{r_0, r_2\}$	14.63	33.09	47.72
$\mathcal{R}_7 = \{r_0, r_3\}$	15.22	26.07	41.29
$\mathcal{R}_8 = \{r_1, r_2\}$	22.30	37.82	60.12
$\mathcal{R}_9 = \{r_1, r_3\}$	26.24	30.03	56.27
$\mathcal{R}_{10} = \{r_2, r_3\}$	21.57	30.63	52.19
$\mathcal{R}_{11} = \{r_0, r_1, r_2\}$	16.52	40.53	57.04
$\mathcal{R}_{12} = \{r_0, r_1, r_3\}$	17.15	32.39	49.54
$\mathcal{R}_{13} = \{r_0, r_2, r_3\}$	16.28	32.97	49.25
$\mathcal{R}_{14} = \{r_1, r_2, r_3\}$	28.49	35.96	64.45
$\mathcal{R}_{15} = \{r_0, r_1, r_2, r_3\}$	18.22	37.77	55.99

A. Simulation Settings

Consider the network topology in Fig. 4 where there are two sessions (s_0, d_0) and (s_1, d_1) . There are four relay nodes that can be employed, namely, r_0, r_1, r_2 and r_3 . We assume that the channel bandwidth in the network is $W = 20$ MHz, the white Gaussian noise at all the nodes has a variance of 10^{-10} W, and every node transmits at a power of 1 W. We assume the channel gain between two nodes u and v is $\|u - v\|^{-4}$, where $\|u - v\|$ is the distance (in meters) between u and v , the path loss index is 4.

B. Results

We calculate the achievable rate of each session under NC-CC when different set of relay nodes are used. Since there are four relay nodes, there are 16 possibilities (see first column in Table II), ranging from the trivial case of no relay node is used to the extreme case when all four relay nodes are used. Table II lists the data rates for each session under each of these 16 cases. When $\mathcal{R} = \emptyset$, Eq. (2) (for direct transmission) is used and the effective bandwidth available to each session is $\frac{W}{2} = 10$ MHz. The last column in Table II shows the sum of the data rates of both sessions.

For the 15 sets of relay nodes (excluding direct transmission), we observe that not every set is beneficial to both sessions, compared to the case of direct transmission. For session (s_0, d_0) , the sets of $\mathcal{R}_1, \mathcal{R}_3, \mathcal{R}_5, \mathcal{R}_6, \mathcal{R}_7, \mathcal{R}_{11}$, and \mathcal{R}_{13} are not beneficial, whereas for session (s_1, d_1) , the sets of \mathcal{R}_1 and \mathcal{R}_4 are not beneficial. Some sets of relay nodes are beneficial to one session but not to the other session, e.g.,

relay groups $\mathcal{R}_3, \mathcal{R}_5, \mathcal{R}_6, \mathcal{R}_7, \mathcal{R}_{11}$, and \mathcal{R}_{13} are beneficial to (s_1, d_1) but not to (s_0, d_0) . Also, the most beneficial set of relay nodes for (s_0, d_0) is \mathcal{R}_{14} , and the most beneficial set of relay nodes for (s_1, d_1) is \mathcal{R}_{11} .

Furthermore, as per our discussion at the end of Section III, we can observe that \mathcal{R}_4 is beneficial to session (s_0, d_0) but not to session (s_1, d_1) due to its proximity to source s_0 . Similarly, \mathcal{R}_3 is beneficial to (s_1, d_1) but not to (s_0, d_0) due to its proximity to s_1 compared to s_0 . For some other relay groups, the conclusion may not be made simply by observing the location of individual nodes. This is where our derived equations can help in determining whether a session group should be matched with a relay group or not. It is important to realize that without deriving the mathematical equations for achievable data rate and mutual information, it is impossible to construct a table such as the one shown in Table II. The formulas we developed can thus help us to identify the optimal set of relay nodes for some particular objective (e.g., maximum sum rate or max-min). The design of such an algorithm, however, is beyond the scope of this paper.

V. CONCLUSION

In this paper, we studied how NC-CC behaves when multiple relay nodes are employed. We derived closed form equations for mutual information and achievable data rate for each session under such setting. Our derivation was based on generalizing the three-node relay channel model with multiple relays and then taking consideration of multiple sessions. Our results in this paper offer an important building block on the theory of NC-CC. To demonstrate the application of our theoretical result, we applied it in a numerical study to understand the impact on a session's achievable rate when different sets of relay nodes are employed.

The results of this paper pave the way for the research community to explore new problems such as finding an optimal set of relay nodes for a given objective. This is our current research and its results will be reported in a future paper.

ACKNOWLEDGMENTS

The work of Y.T. Hou, S. Sharma, and Y. Shi was supported in part by the NSF under Grant CCF-0946273. The work of S. Kompella was supported in part by the ONR.

REFERENCES

[1] B. Aazhang, R.S. Blum, J.N. Laneman, K.J.R. Liu, W. Su, and A. Wittneben (editors), *IEEE Journal on Selected Areas in Com-*

munications – Special Issue on Cooperative Communications and Networking, vol. 25, no. 2, Feb. 2007.

[2] X. Bao and J. Li, “Adaptive network coded cooperation (ANCC) for wireless relay networks: matching code-on-graph with network-on-graph,” *IEEE Transactions on Wireless Communications*, vol. 7, no. 2, pp. 574–583, February 2008.

[3] F.H.P. Fitzek, M. Dohler, I. Opperman, A. Molinaro, M. Katz, and C. Bettstetter (editors), *Springer Journal on Wireless Personal Communications – Special Issue on Cooperation in Wireless Networks*, vol. 43, no. 1, January 2007.

[4] S. Katti, S. Gollakotta, and D. Katabi, “Embracing wireless interference: Analog network coding,” in *Proc. ACM SIGCOMM*, pp. 397–408, Kyoto, Japan, August 27–31, 2007.

[5] G. Kramer, I. Maric, and R.D. Yates, “Cooperative communications,” *Foundations and Trends in Networking*, vol. 1, issue 3–4, Now Publishers, June 2007.

[6] G. Kramer, R. Berry, A. El Gamal, H. El Gamal, M. Franceschetti, M. Gastpar, and J.N. Laneman (editors), *IEEE Transactions on Information Theory – Special Issue on Models, Theory, and Codes for Relaying and Cooperation in Communication Networks*, vol. 53, no. 10, Oct. 2007.

[7] J.N. Laneman, D.N.C. Tse, and G.W. Wornell, “Cooperative diversity in wireless networks: Efficient protocols and outage behavior,” *IEEE Transactions on Information Theory*, vol. 50, no. 12, pp. 3062–3080, Dec. 2004.

[8] L.E. Li, R. Alimi, D. Shen, H. Vishwanathan, and Y.R. Yang, “A general algorithm for interference alignment and cancellation in wireless networks,” in *Proc. IEEE INFOCOM*, San Diego, CA March 15–19, 2010.

[9] C. Peng, Q. Zhang, M. Zhao, and Y. Yao, “On the performance analysis of network-coded cooperation in wireless networks,” in *Proc. IEEE INFOCOM*, pp. 1460–1468, Anchorage, Alaska, May 6–12, 2007.

[10] S. Sharma, Y. Shi, J. Liu, and Y.T. Hou, “Is network coding always good for cooperative communications,” in *Proc. IEEE INFOCOM*, San Diego, CA, March 15–19, 2010.

[11] Y. Shi, S. Sharma, Y.T. Hou, and S. Kompella, “Optimal relay assignment for cooperative communications,” in *Proc. ACM MobiHoc*, pp. 3–12, Hong Kong, China, May 27–30, 2008.

[12] Y. Tu and G.J. Pottie, “Coherent cooperative transmission from multiple adjacent antennas to a distant stationary antenna through AWGN channels,” in *Proc. IEEE Vehicular Technology Conference*, pp. 130–134, Birmingham, AL, May 6–9, 2002.

[13] L. Xiao, T.E. Fuja, J. Kliewer, and D.J. Costello, “A network coding approach to cooperative diversity,” *IEEE Transactions on Information Theory*, vol. 53, no. 10, pp. 3714–3722, Oct. 2007.

[14] H. Xu and B. Li, “XOR-assisted cooperative diversity in OFDMA wireless networks: Optimization framework and approximation algorithms,” in *Proc. IEEE INFOCOM*, pp. 2141–2149, Rio de Janeiro, Brazil, April 19–25, 2009.